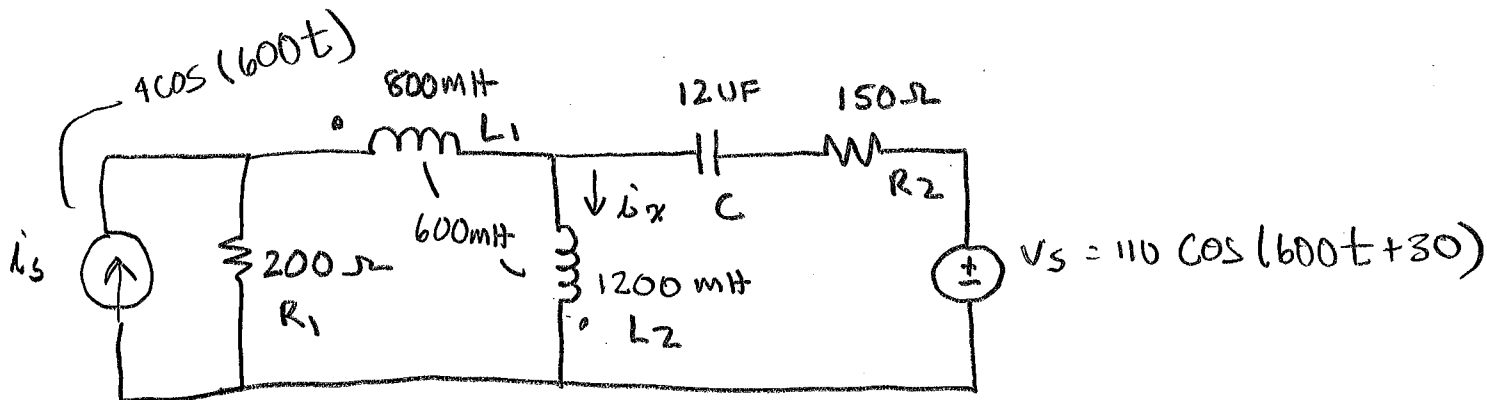


LECTURE 15 - FUN WITH TRANSFORMERS

11

(3.11) Find i_x Using mesh analysis

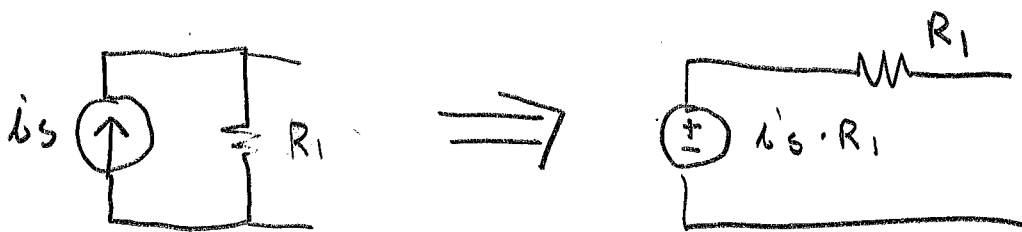


The coils are coupling, but a lot or just a bit?

We know $M \leq \sqrt{L_1 L_2}$ EQ 13.35

and $K = \frac{M}{\sqrt{L_1 L_2}} = \frac{0.6}{\sqrt{0.8 \times 1.2}} = 0.6124$

for mesh analysis we'll use source transformation



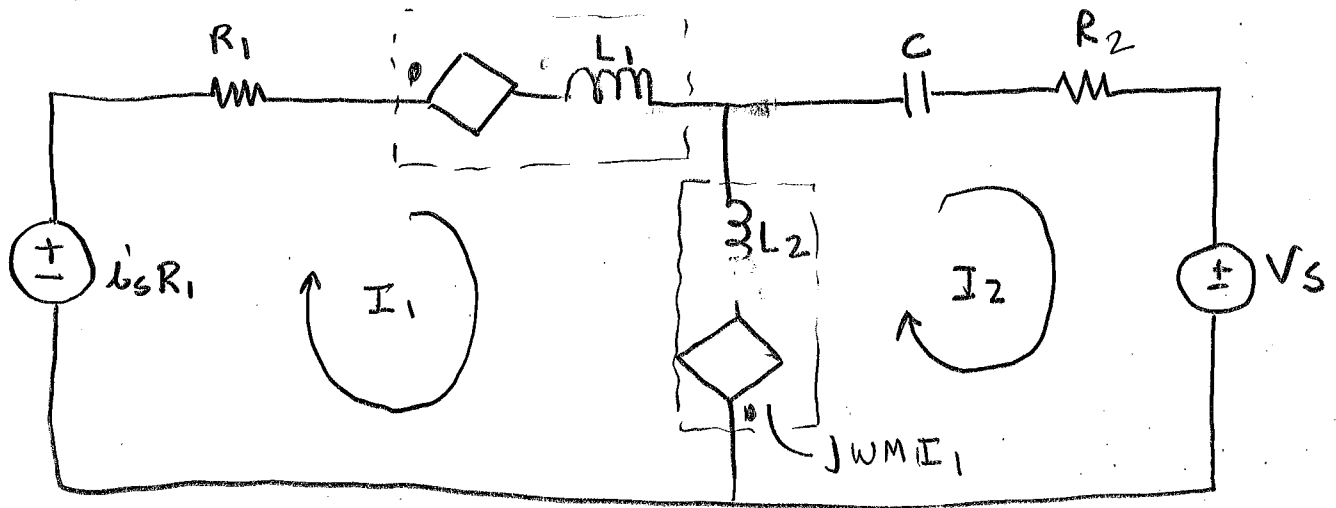
$i_s = 4 \cos(600t) \rightarrow$ what is Freq in Hz?

- a - 600 Hz
- b - 3769 Hz
- c - none of the above

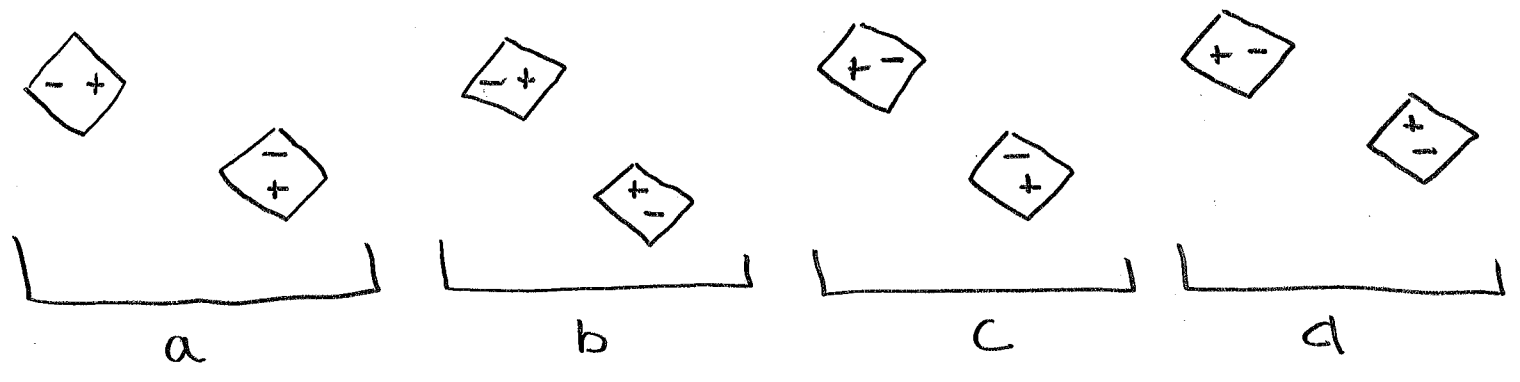
PR 13.11 CONT

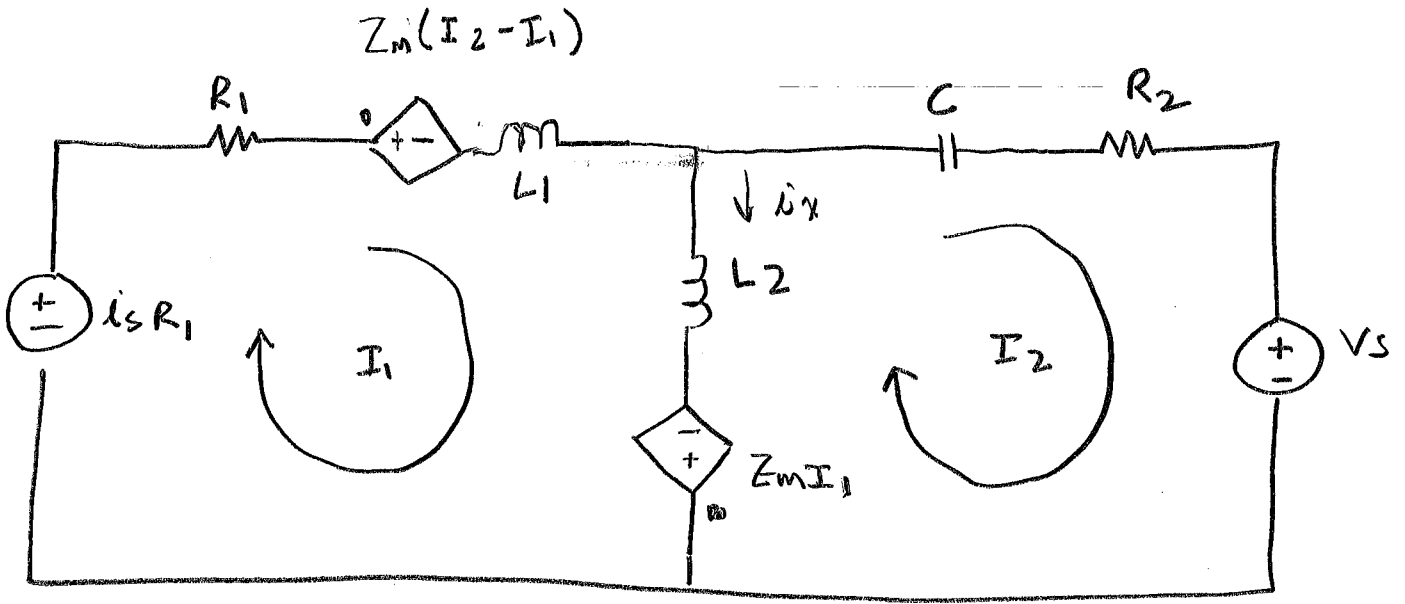
12

$j\omega M(I_2 - I_1)$



"It is... time to vote"





$$\textcircled{1} -I_s R_1 + I_1 R_1 + Z_m(I_2 - I_1) + Z_{L1} I_1 + Z_{L2}(I_1 - I_2) - Z_m I_1 = 0$$

$$I_1 (R_1 - Z_m + Z_{L1} + Z_{L2} - Z_m) + I_2 (Z_m - Z_{L2}) = I_s R_1$$

$$\textcircled{2} Z_m I_1 + (I_2 - I_1) Z_{L2} + I_2 Z_C + I_2 R_2 + V_s = 0$$

$$I_1 (Z_m - Z_{L2}) + I_2 (Z_{L2} + Z_C + R_2) = -V_s$$

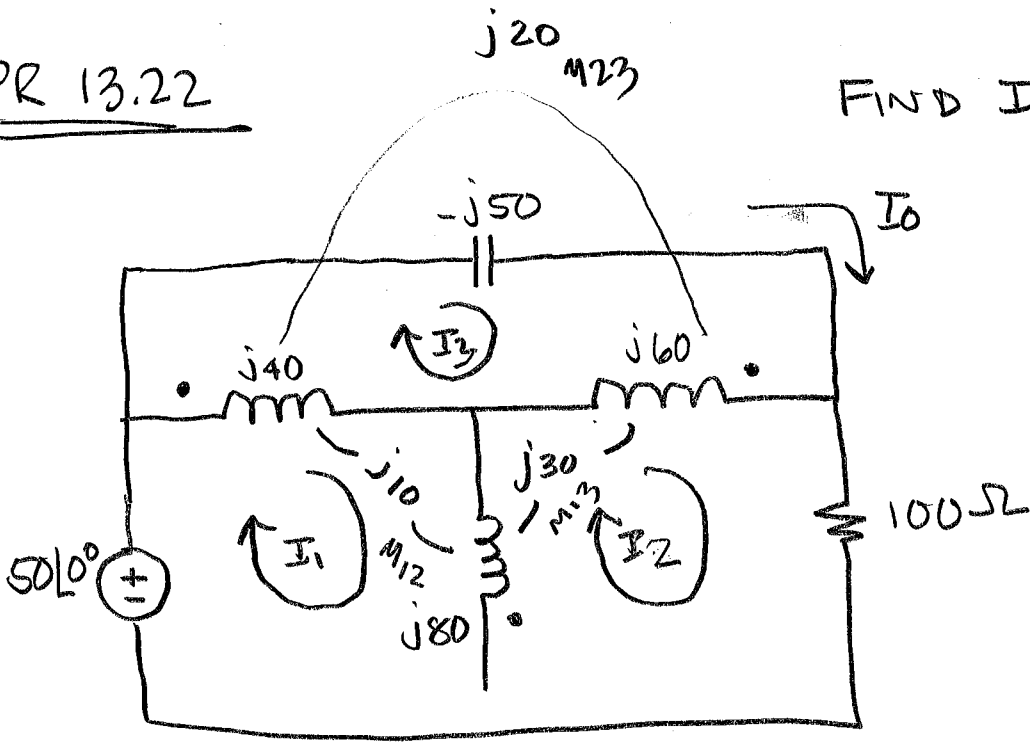
$$\begin{bmatrix} R_1 - Z_m + Z_{L1} + Z_{L2} - Z_m & Z_m - Z_{L2} \\ Z_m - Z_{L2} & Z_{L2} + Z_C + R_2 \end{bmatrix}^{-1} \begin{bmatrix} I_s R_1 \\ -V_s \end{bmatrix} = \begin{bmatrix} I_1 \\ I_2 \end{bmatrix}$$

and $I_x = I_1 - I_2 = 1.07 \cos(600t - 66^\circ)$

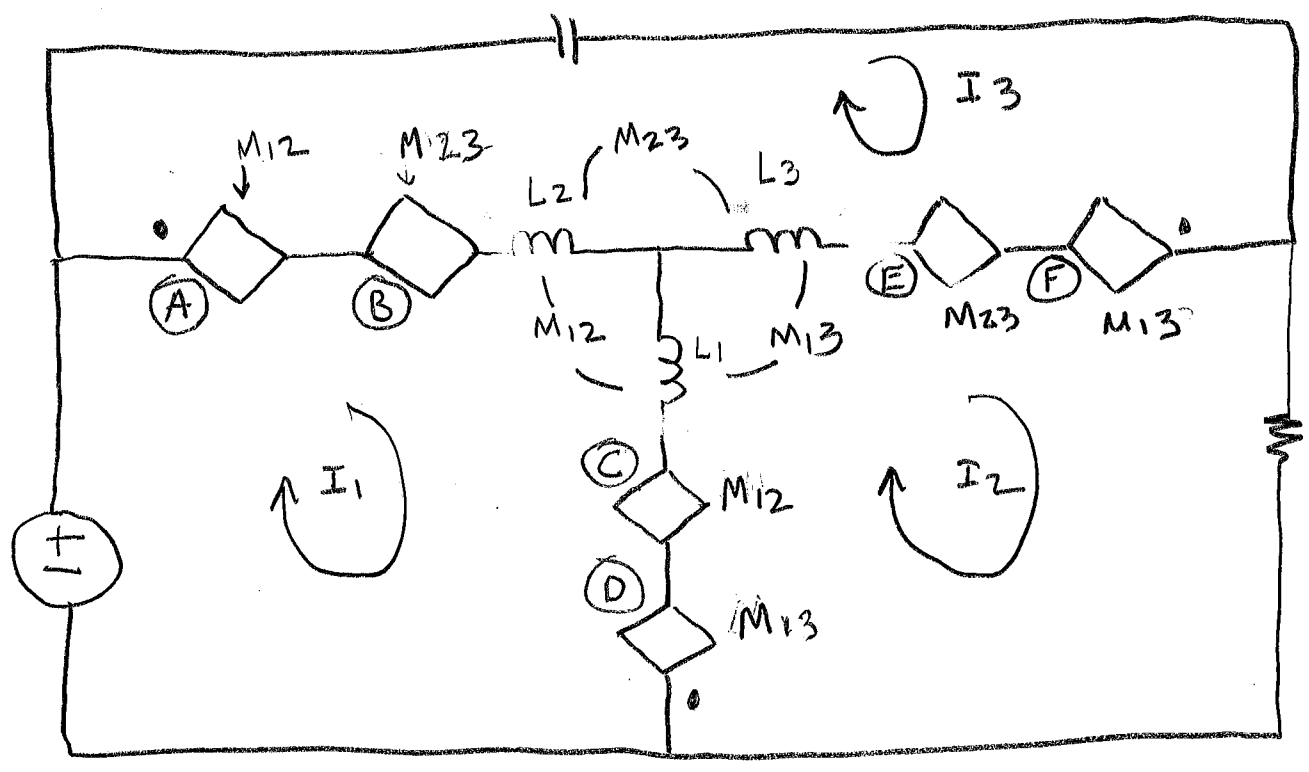
I disagree with answer in back of book

PR 13.22

FIND I_0



"Bridged-T" network



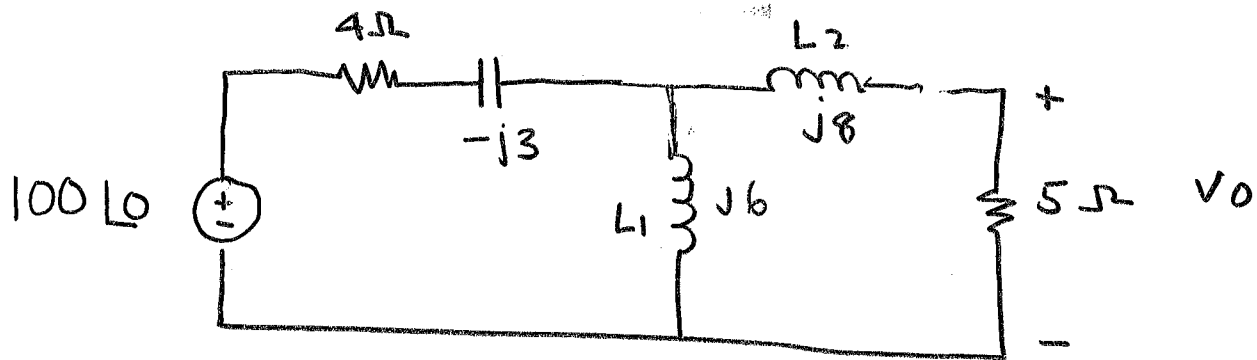
made current entering dot of "donor" positive

- (A) → $S M_{12} (I_2 - I_1)$
- (B) → $S M_{23} (I_3 - I_2)$
- (C) → $S M_{12} (I_1 - I_3)$
- (D) → $S M_{13} (I_3 - I_2)$
- (E) → $S M_{23} (I_1 - I_3)$
- (F) → $S M_{13} (I_2 - I_1)$

FILL IN THE POLARITY ON THE SOURCES

Practical problem

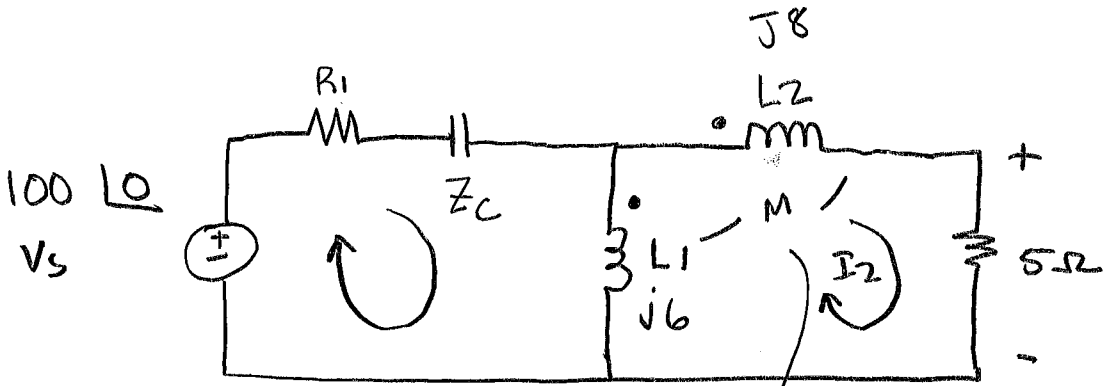
16



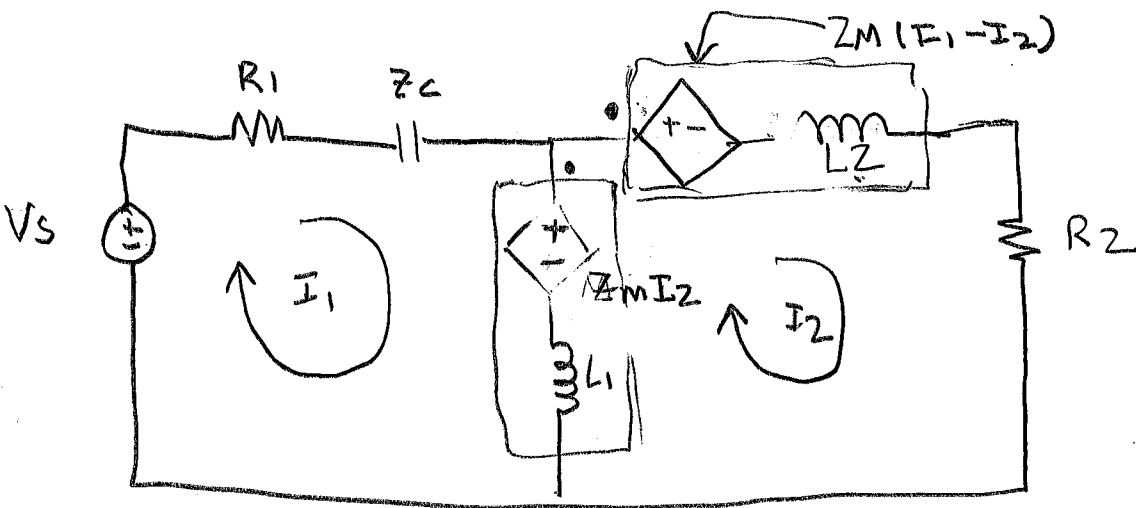
- 1) "There is no mutual inductance." Compute V_0 (design value)
- 2) Measure the output voltage, see if it equals the design value.
- 4) IF the output voltage doesn't match the design value, compute the mutual inductance.

Strategy

- 1) Look at the circuit. Hypothesize where dots are
- 2) Compute the expected output with $M = 0$. This is the design value.
- 3) Measure the output voltage
- 4) Adjust M in equations until we match the measured voltage



hypothesized M and dots



I_2 enters dot of L_2 so it makes dot of L_1 more positive

I_1 enters dot of L_1 so it makes dot of L_2 more positive

note $(I_1 - I_2)Z_M$ for L_2

$$\textcircled{1} -V_s + R_1 I_1 + Z_c I_1 + Z_M I_2 + (I_1 - I_2) Z_{L1} = 0$$

$$I_1 (R_1 + Z_c + Z_{L1}) + I_2 (+Z_M - Z_{L1}) = V_s$$

$$\textcircled{2} (I_2 - I_1) Z_{L1} - Z_M I_2 + I_2 Z_{L2} + Z_M (I_1 - I_2) + I_2 R_2 = 0$$

$$I_1 (-Z_{L1} + Z_M) + I_2 (Z_{L1} - Z_M + Z_{L2} - Z_M + R_2) = 0$$

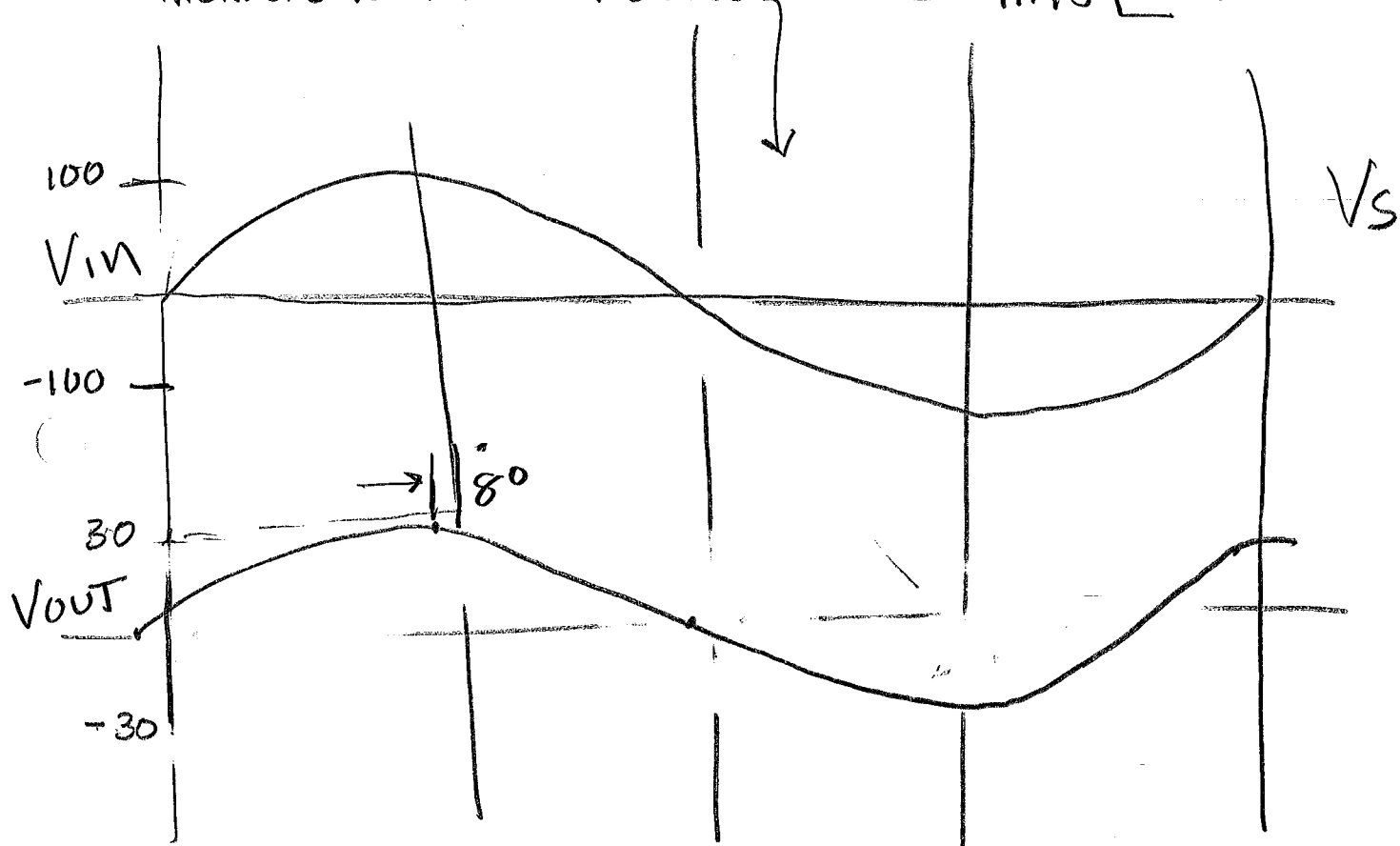
$$\begin{bmatrix} R_1 + Z_{L1} + Z_{L2} & +Z_M - Z_{L1} \\ -Z_{L1} + Z_M & Z_{L1} - Z_M + Z_{L2} - Z_M + R_2 \end{bmatrix}^{-1} \begin{bmatrix} V_s \\ 0 \end{bmatrix} = \begin{bmatrix} I_1 \\ I_2 \end{bmatrix}$$

WITH $Z_M = 0$, NO COUPLING

$$V_o = I_2 R_2 = (8.134 + j1.60) \times 5 = 40.67 + j8.02$$

$$\begin{matrix} \downarrow & \downarrow \\ 41.45 & \angle 11.150 \end{matrix}$$

MEASURE WAVEFORM BELOW



Is the circuit working right?

So what do we do?

1) remove the components and measure them
- They are ok

2) Hypothesize mutual coupling.

- Note desired phase was 11.15° , we got about 8°
That's pretty hard to measure.

- But we expected amplitude of 41 and we got about 27. That's easy to see.

3)

ZM	Vol	L phase	
j1	39.3	8.1°	
j2	36.1	6.2°	
j3	31.7	6.07°	
j4	25.34	8.74°	← Too far
j3.77	27.04	7.79	← matches what we saw

can I do this?

$$K = \text{Coefficient of coupling} = \frac{M}{\sqrt{L_1 L_2}} \stackrel{?}{=} \frac{j\omega M}{\sqrt{j\omega L_1 \cdot j\omega L_2}}$$

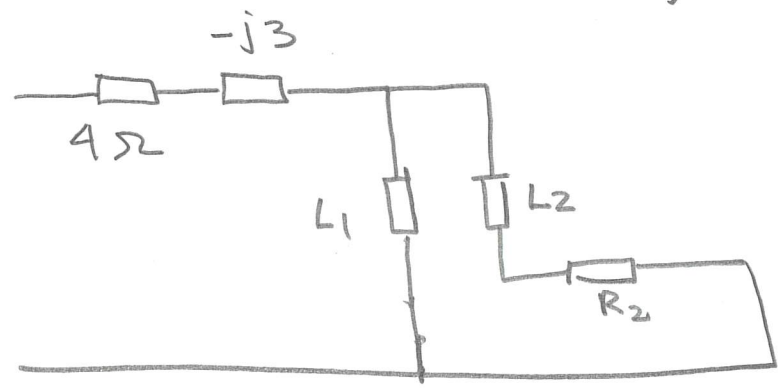
yes

$$= \frac{j\omega M}{\sqrt{-1\omega^2 L_1 L_2}} = \frac{j\omega M}{j\omega \sqrt{L_1 L_2}} = \frac{M}{\sqrt{L_1 L_2}}$$

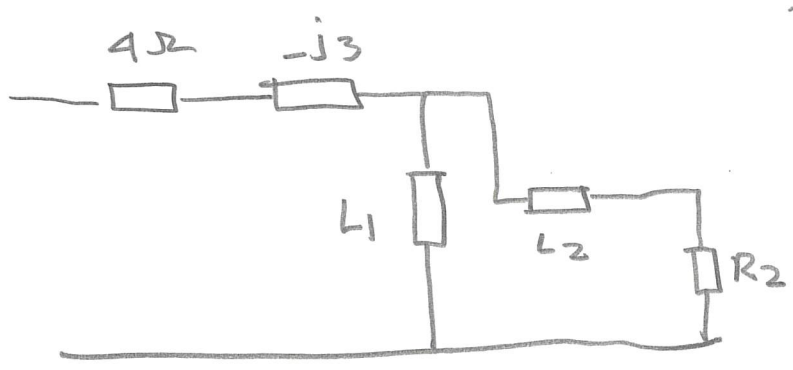
$$\text{So } K = \frac{j3.77}{\sqrt{j6 \times j8}} = 0.5442$$



Found this layout

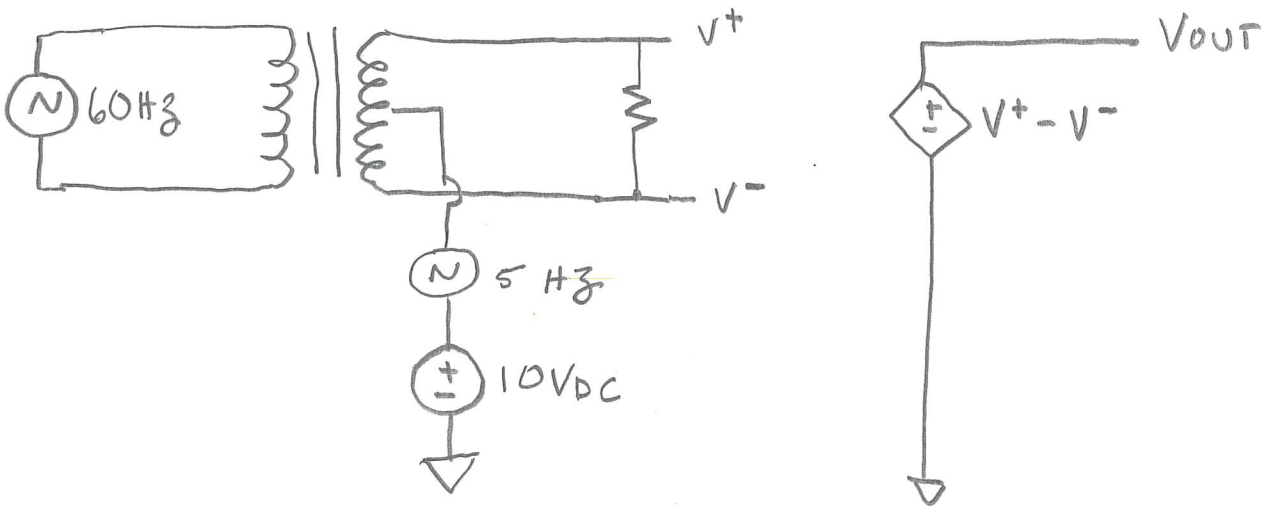


What will this do?



Fixed it!

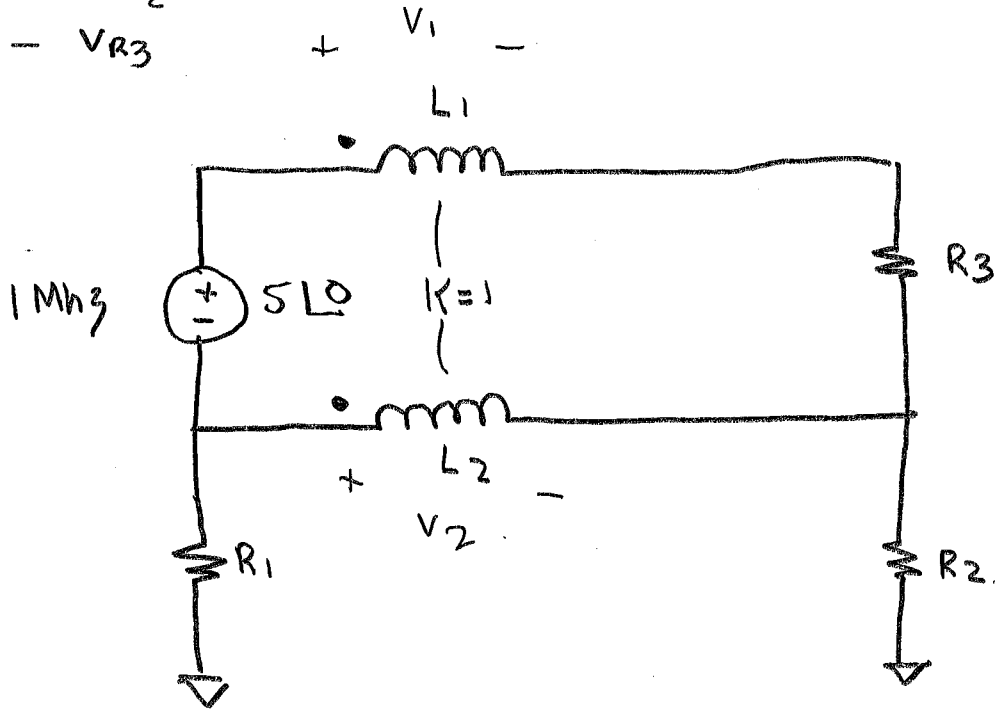
Center Tapped Transformer

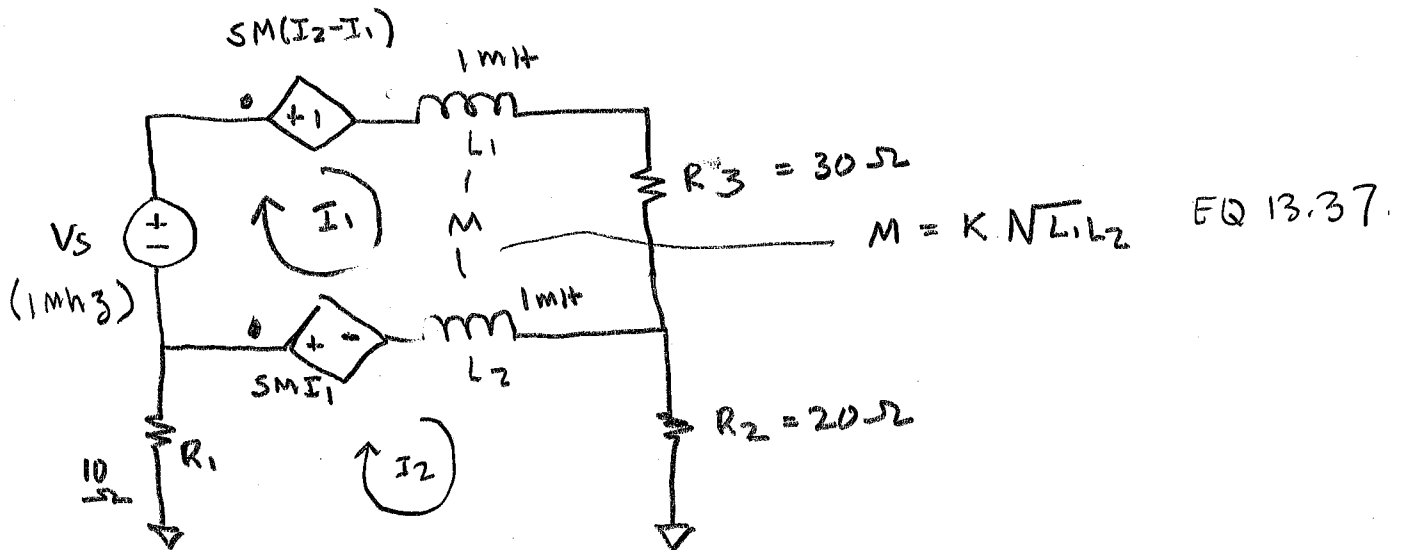


LONGITUDINAL TRANSFORMERS

FIND

- V_1
- V_2
- I_{R_1}
- I_{R_2}
- V_{R_3}





$$\textcircled{1} -V_s + SM(I_2 - I_1) + SL_1 I_1 + R_3 I_1 + SL_2(I_1 - I_2) - SM I_1$$

$$\textcircled{1} -V_s + SM I_2 - SM I_1 + SL_1 I_1 + I_1 R_3 + SL_2 I_1 - SL_2 I_2 - SM I_1 = 0$$

$$\textcircled{1} (-SM + SL_1 + R_3 + SL_2 - SM) I_1 + (SM - SL_2) I_2 = V_s$$

$$\textcircled{2} I_2 R_1 + SM I_1 + SL_2(I_2 - I_1) + I_2 R_2 = 0$$

$$\textcircled{2} (SM - SL_2) I_1 + (R_1 + SL_2 + R_2) I_2 = 0$$

Substitute values - $S = j2\pi \times 1\text{mH}3$, $R_1 = 10 \Omega$, $R_2 = 20 \Omega$, $R_3 = 30 \Omega$
 $L_1 = L_2 = 1\text{mH}$

$$\begin{bmatrix} -SM + SL_1 + R_3 + SL_2 - SM & SM - SL_2 \\ SM - SL_2 & R_1 + SL_2 + R_2 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} V_s \\ 0 \end{bmatrix}$$

Say $L_1 = L_2 = 1\text{mH}$, $K = 1$ Then $I_1 = 167 \text{mA}$, $I_2 = 0$

$$V_{L1} = SM(I_2 - I_1) + SL_1 I_1 = 0$$

$$V_{L2} = SM I_1 + SL_2(I_2 - I_1) = 0$$

$$V_{R3} = R_3 \times I_1 = 5V \rightarrow \text{Hey, thats } V_s$$

$$V_{R2} = I_2 R_2 = 0, V_{R1} = I_2 R_1 = 0$$

in the previous example we note that

1) $k = 1$

2) Impedance of inductors $= j2\pi fL = j6.28 \text{ K}\Omega$
is much larger than the resistances.

Can we use ideal transformer model?

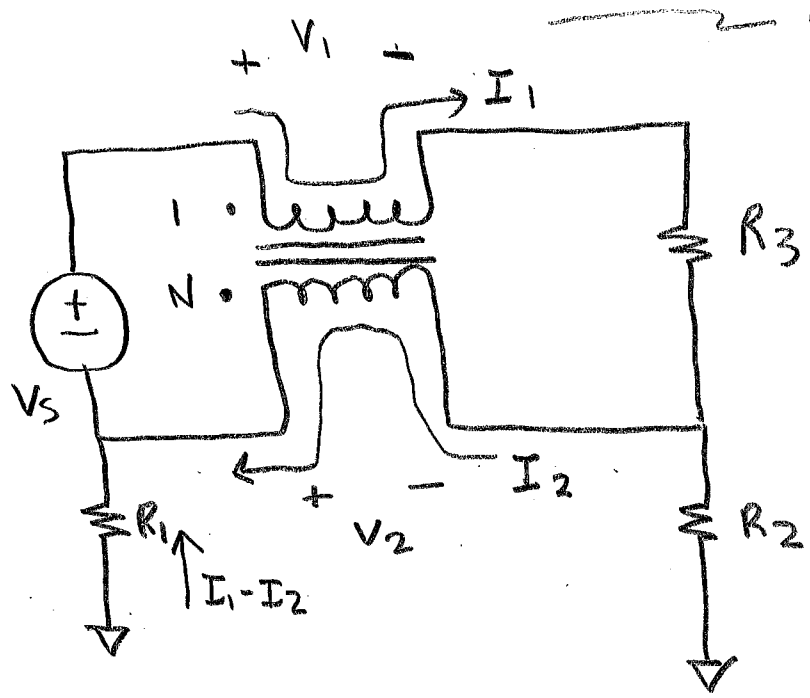
We have L_1 and L_2 , what is turns ratio N ?

see top of pg 574

$$n = \sqrt{\frac{L_2}{L_1}} = \sqrt{\frac{1 \text{ mH}}{1 \text{ mH}}} = 1.0$$



ideal voltage/current convention from figure 13.31



① upper mesh

$$-V_s + V_1 + I_1 R_3 - V_2 = 0$$

② lower mesh

$$(I_1 - I_2) R_1 + V_2 + (I_1 - I_2) R_2 = 0$$

} 2 EQ 4 unknowns

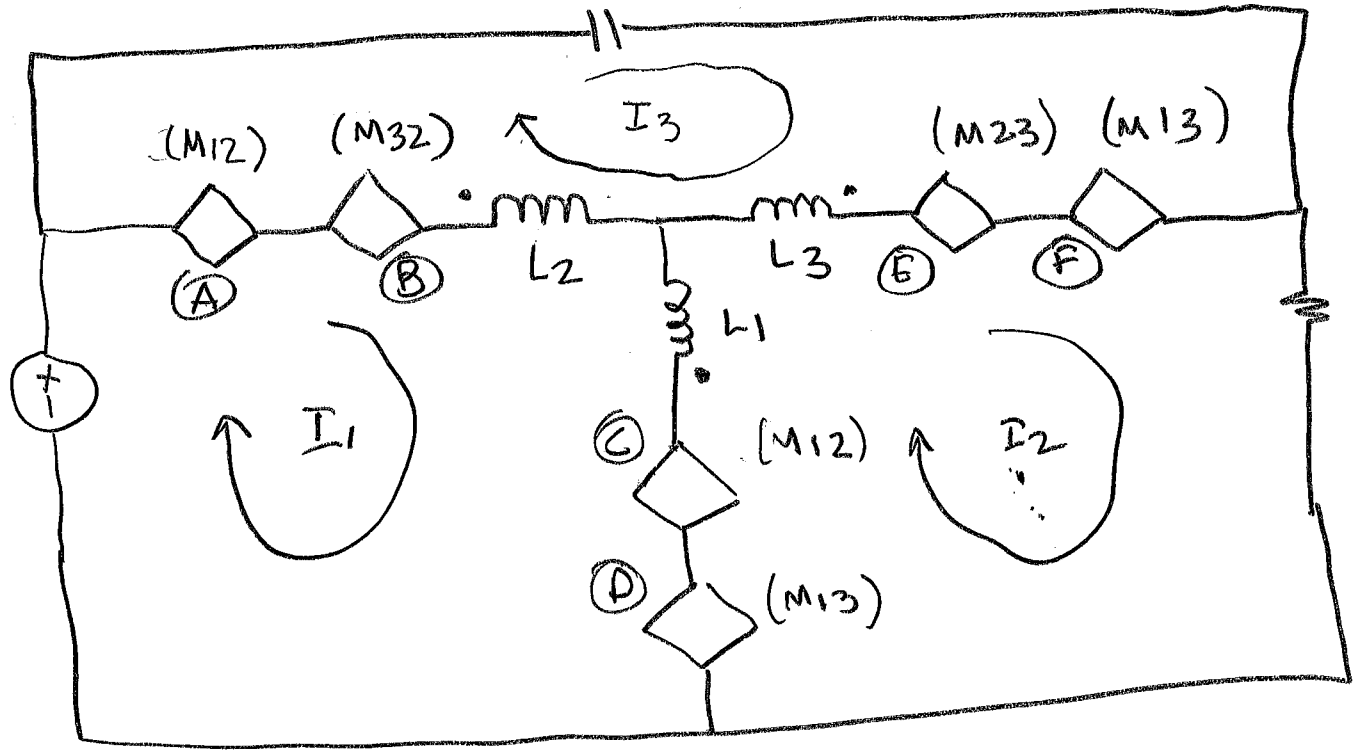
Ideal transformer: $V_2 = N V_1$, $I_2 = I_1 / N$

Since $N = 1$, $I_1 = I_2$

Therefore by KCL, $I_{R1} = I_{R2} = 0$

This is called a "Common Mode choice"

V_s is seen at R_3 Regardless of the values of R_1 and R_2 !



I define current so it enters dot

(A) → $I_{L1} = I_2 - I_1$, I_2 makes dot of L_2 more POS

$\diamond^{+-} SM_{12}(I_2 - I_1)$

(B) → $I_{L3} = I_3 - I_2$, I_3 makes dot of L_2 more positive

$\diamond^{+-} SM_{32}(I_3 - I_2)$

(C) → $I_{L2} = I_1 - I_3$, I_1 makes dot of L_1 more POS

$\diamond^{-+} SM_{12}(I_1 - I_3)$

(D) $I_{L3} = I_3 - I_2$ I_3 makes dot of L_1 more pos

so

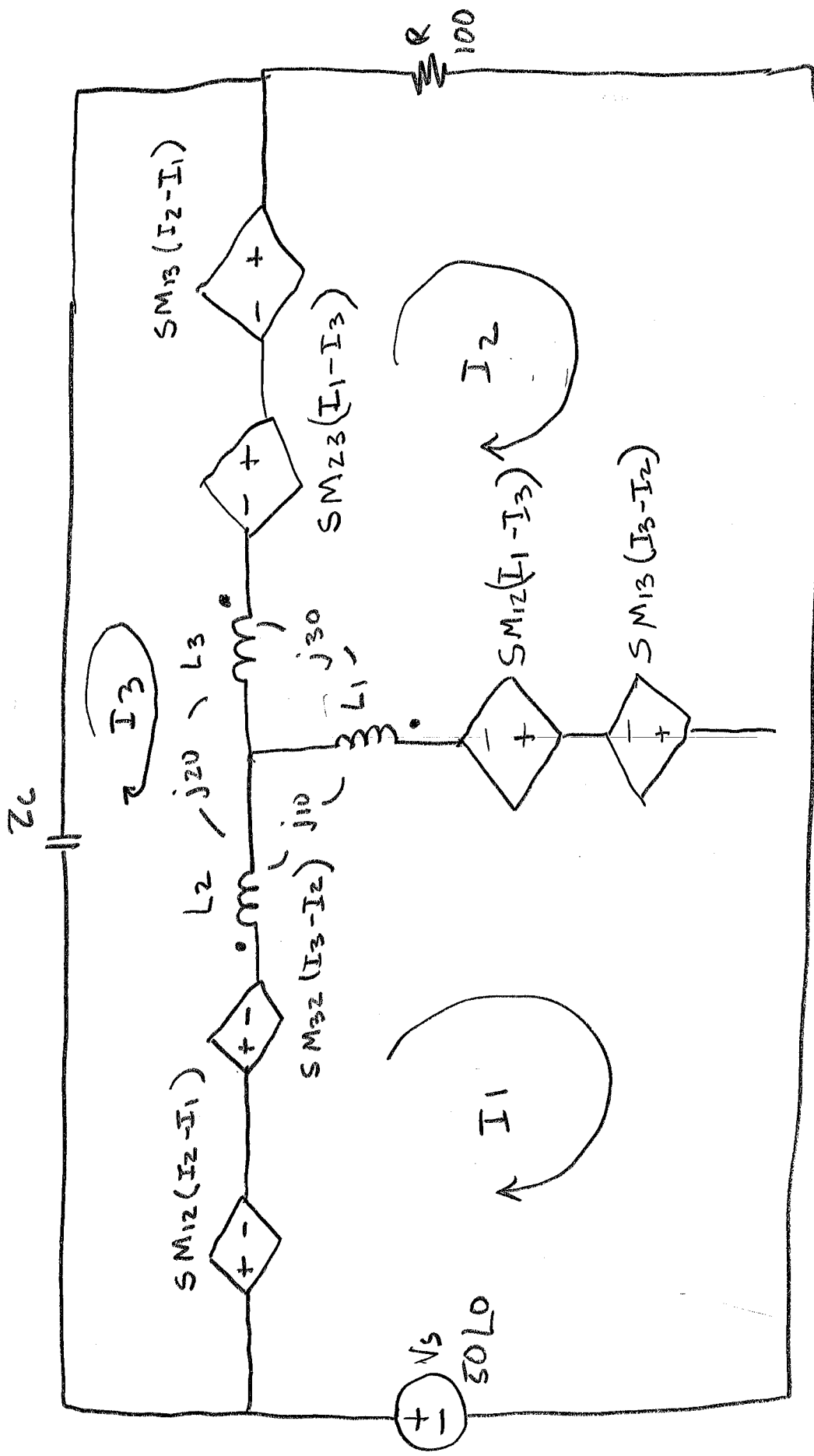
\diamond_{-+} $SM_{13}(I_3 - I_2)$

(E) $I_{L2} = I_1 - I_3$, I_1 makes dot of L_3 more pos

\diamond_{-+} $SM_{23}(I_1 - I_3)$

(F) $I_{L1} = I_2 - I_1$, I_2 makes dot of L_3 more pos

\diamond_{-+} $SM_{13}(I_2 - I_1)$



①

$$-V_s + Z_{12}(I_2 - I_1) + Z_{32}(I_3 - I_2) + Z_{L2}(I_1 + Z_{L1}(I_1 - I_2)) - Z_{12}I_1 - Z_{12}(I_3 - I_2) = 0$$

$$I_1(-Z_{12} + Z_{L2} + Z_{L1} - Z_{12}) + I_2(Z_{12} - Z_{32} - Z_{L1} + Z_{12}) + I_3(Z_{32} - Z_{12}) = V_s$$

② $Z_{13}(I_3 - I_2) + Z_{12}I_1 + Z_{L1}(I_2 - I_1) + Z_{L3}(I_2 - I_3)$

$$-Z_{23}I_1 - Z_{13}(I_2 - I_1) + I_2R = 0 + Z_{23}I_3$$

$$I_1(Z_{12} - Z_{L1} - Z_{23} + Z_{13}) + I_2(-Z_{13} + Z_{L1} + Z_{L3} - Z_{13} + R) + I_3(Z_{13} - Z_{L3}) = 0$$

③ $(I_3 - I_1)Z_{L2} - Z_{32}(I_3 - I_2) - Z_{12}(I_2 - I_1) + I_3Z_C + Z_{13}(I_2 - I_1) + Z_{23}I_1 + (I_3 - I_2)Z_{L3} = 0$

$$I_1(-Z_{L2} + Z_{12} - Z_{13} + Z_{23}) + I_2(Z_{32} - Z_{12} + Z_{13} - Z_{L3}) + I_3(Z_{L2} - Z_{32} + Z_C + Z_{L3}) = 0$$

SOLVING Gives

$$I_1 = 0.1885 + j0.143$$

$$I_2 = -0.0808 + j0.2962$$

$$I_3 = 0.5923 + j1.1615 = I_0 = 1.30 \angle 62.98^\circ$$